Outline

- Circular and linear convolutions
- 2D DFT
- 2D DCT
- Properties
- Other formulations
- Examples

Circular convolution

- Finite length signals (N_0 samples) \rightarrow circular or periodic convolution
 - the summation is over 1 period
 - the result is a N₀ period sequence

$$c[k] = f[k] \otimes g[k] = \sum_{n=0}^{N_0 - 1} f[n]g[k - n]$$

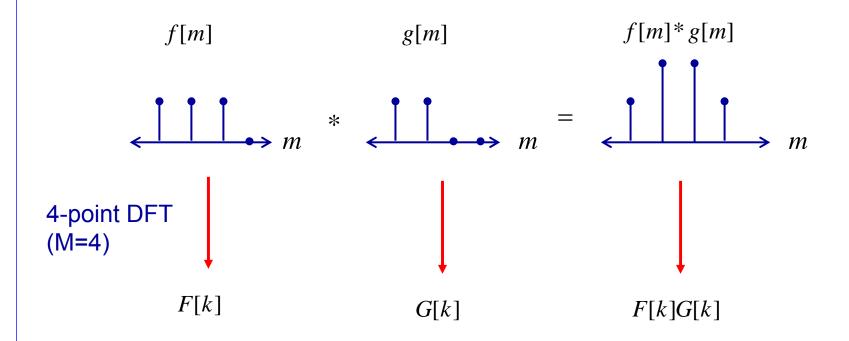
 The circular convolution is equivalent to the linear convolution of the zero-padded equal length sequences

For the convolution property to hold, M must be *greater than or equal* to P+Q-1.

Convolution

$$f[m] * g[m] \Leftrightarrow F[k]G[k]$$

Zero padding



In words

Given 2 sequences of length N and M, let y[k] be their linear convolution

$$y[k] = f[k] * h[k] = \sum_{n=-\infty}^{+\infty} f[n]h[k-n]$$

 y[k] is also equal to the circular convolution of the two suitably zero padded sequences making them consist of the same number of samples

$$c[k] = f[k] \otimes h[k] = \sum_{n=0}^{N_0 - 1} f[n]h[k - n]$$

$$N_0 = N_f + N_h - 1$$
: length of the zero-padded seq

- In this way, the linear convolution between two sequences having a different length (filtering) can be computed by the DFT (which rests on the circular convolution)
 - The procedure is the following
 - Pad f[n] with N_h-1 zeros and h[n] with N_f-1 zeros
 - Find Y[r] as the product of F[r] and H[r] (which are the DFTs of the corresponding zero-padded signals)
 - Find the inverse DFT of Y[r]
- Allows to perform linear filtering using DFT

 Fourier transform of a 2D signal defined over a discrete finite 2D grid of size MxN

or equivalently

- Fourier transform of a 2D set of samples forming a bidimensional sequence
- As in the 1D case, 2D-DFT, though a self-consistent transform, can be considered as a mean of calculating the transform of a 2D sampled signal defined over a discrete grid.
- The signal is periodized along both dimensions and the 2D-DFT can be regarded as a sampled version of the 2D DTFT

2D Discrete Fourier Transform (2D DFT)

2D Fourier (discrete time) Transform (DTFT) [Gonzalez]

$$F(u,v) = \sum_{m=-\infty}^{\infty} \sum_{n=-\infty}^{\infty} f[m,n]e^{-j2\pi(um+vn)}$$
 a-periodic signal periodic transform

• 2D Discrete Fourier Transform (DFT)

$$F[k,l] = \frac{1}{MN} \sum_{m=0}^{M-1} \sum_{n=0}^{N-1} f[m,n] e^{-j2\pi \left(\frac{k}{M}m + \frac{l}{N}n\right)}$$

periodized signal periodic and sampled transform

2D DFT can be regarded as a sampled version of 2D DTFT.

2D DFT: Periodicity

- A [M,N] point DFT is periodic with period [M,N]
 - Proof

$$F[k,l] = \frac{1}{MN} \sum_{m=0}^{M-1} \sum_{n=0}^{N-1} f[m,n] e^{-j2\pi \left(\frac{k}{M}m + \frac{l}{N}n\right)}$$

$$F[k+M,l+N] = \frac{1}{MN} \sum_{m=0}^{M-1} \sum_{n=0}^{N-1} f[m,n] e^{-j2\pi \left(\frac{k+M}{M}m + \frac{l+N}{N}n\right)}$$

$$= \frac{1}{MN} \sum_{m=0}^{M-1} \sum_{n=0}^{N-1} f[m,n] e^{-j2\pi \left(\frac{k}{M}m + \frac{l}{N}n\right)} e^{-j2\pi \left(\frac{M}{M}m + \frac{N}{N}n\right)}$$

$$= F[k,l]$$

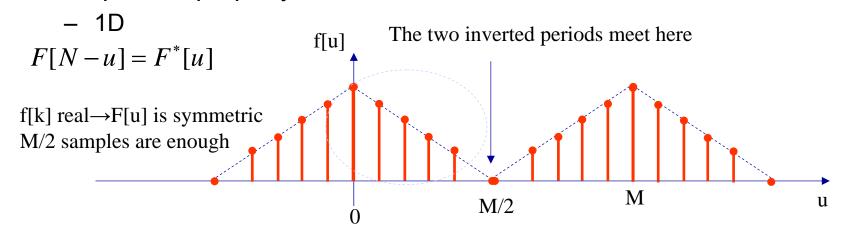
(In what follows: spatial coordinates=k,l, frequency coordinates: u,v)

2D DFT: Periodicity

Periodicity

$$F[u,v] = F[u+mM,v] = F[u,v+nN] = F[u+mM,v+nN]$$
$$f[k,l] = f[k+mM,l] = f[k,l+nN] = f[k+mM,l+nN]$$

 This has important consequences on the implementation and energy compaction property



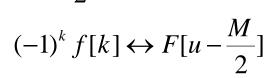
Periodicity: 1D

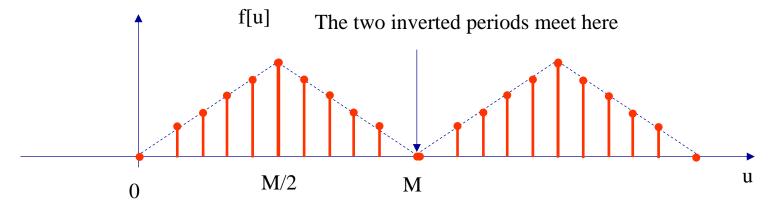
$$f[k] \leftrightarrow F[u]$$

$$f[k]e^{j2\pi\frac{u_0k}{M}} \leftrightarrow F[u-u_0]$$

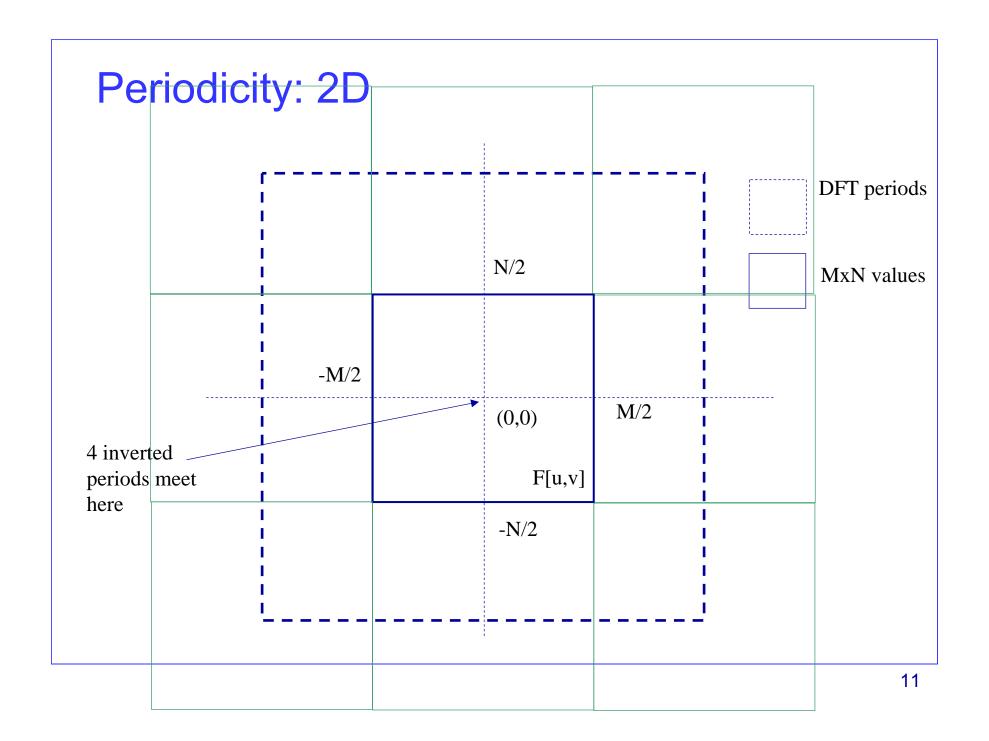
$$u_0 = \frac{M}{2} \to e^{j2\pi\frac{u_0k}{M}} = e^{j2\pi\frac{Mk}{2M}} = e^{j\pi k} = (-1)^k$$

changing the sign of every other sample puts F[0] at the center of the interval [0,M]

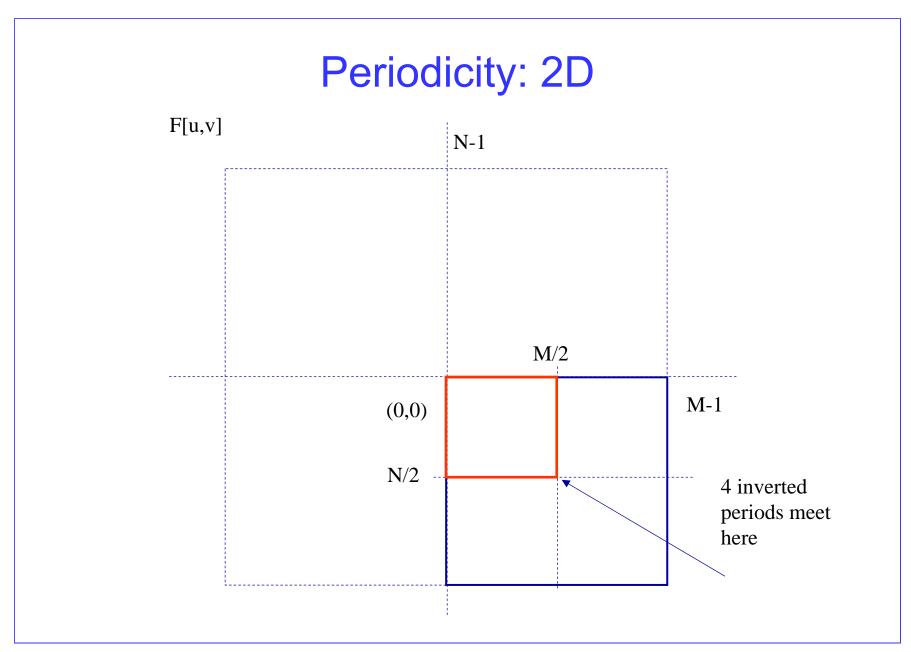




It is more practical to have one complete period positioned in [0, M-1]



Periodicity: 2D data contain one centered F[u,v]complete period N-1 DFT periods $f[k,l]e^{j2\pi(\frac{u_0k}{M}+\frac{v_0l}{N})} \longleftrightarrow F[u-u_0,v-v_0]$ $u_0 = \frac{M}{2}, v_0 = \frac{N}{2} \rightarrow (-1)^{k+l} f[k] \leftrightarrow F \left[u - \frac{M}{2}, v - \frac{N}{2} \right]^{k+l}$ MxN values M/2M-1(0,0)4 inverted periods meet here



Periodicity in spatial domain

[M,N] point inverse DFT is periodic with period [M,N]

$$f[m,n] = \sum_{k=0}^{M-1} \sum_{l=0}^{N-1} F[k,l] e^{j2\pi \left(\frac{k}{M}m + \frac{l}{N}n\right)}$$

$$f[m+M,n+N] = \sum_{k=0}^{M-1} \sum_{l=0}^{N-1} F[k,l] e^{j2\pi \left(\frac{k}{M}(m+M) + \frac{l}{N}(n+N)\right)}$$

$$= \sum_{k=0}^{M-1} \sum_{l=0}^{N-1} F[k,l] e^{j2\pi \left(\frac{k}{M}m + \frac{l}{N}n\right)} e^{j2\pi \left(\frac{k}{M}M + \frac{l}{N}N\right)}$$

$$= f[m,n]$$

Angle and phase spectra

$$F[u,v] = |F[u,v]|e^{j\Phi[u,v]}$$

$$|F[u,v]| = [\operatorname{Re}\{F[u,v]\}^2 + \operatorname{Im}\{F[u,v]\}^2]^{1/2}$$

modulus (amplitude spectrum)

$$\Phi[u,v] = \arctan\left[\frac{\operatorname{Im}\left\{F[u,v]\right\}}{\operatorname{Re}\left\{F[u,v]\right\}}\right]$$

phase

$$P[u,v] = |F[u,v]|^2$$

power spectrum

For a real function

$$F[-u,-v] = F^*[u,v]$$

conjugate symmetric with respect to the origin

$$|F[-u,-v]| = |F[u,v]|$$

$$\Phi[-u,-v] = -\Phi[u,v]$$

Translation and rotation

$$f[k,l]e^{j2\pi\left(\frac{m}{M}k+\frac{n}{N}l\right)} \longleftrightarrow F\left[u-m,v-l\right]$$

$$f\left[k-m,l-n\right] \longleftrightarrow F\left[u,v\right]^{-j2\pi\left(\frac{m}{M}k+\frac{n}{N}l\right)}$$

$$\begin{cases} k = r \cos \theta & \begin{cases} u = \omega \cos \varphi \\ l = r \sin \theta & l = \omega \sin \varphi \end{cases}$$

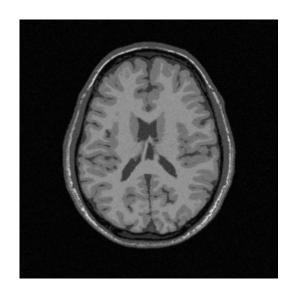
$$f \left[r, \theta + \theta_0 \right] \longleftrightarrow F \left[\omega, \varphi + \theta_0 \right]$$

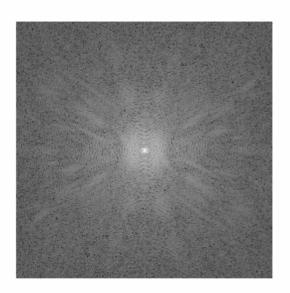
Rotations in spatial domain correspond equal rotations in Fourier domain

mean value

$$F[0,0] = \frac{1}{NM} \sum_{n=0}^{N-1} \sum_{m=0}^{M-1} f[n,m]$$

DC coefficient





Separability

 The discrete two-dimensional Fourier transform of an image array is defined in series form as

$$F[k,l] = \frac{1}{MN} \sum_{m=0}^{M-1} \sum_{n=0}^{N-1} f[m,n] e^{-j2\pi \left(\frac{k}{M}m + \frac{l}{N}n\right)}$$

inverse transform

$$f[m,n] = \sum_{k=0}^{M-1} \sum_{l=0}^{N-1} F[k,l] e^{j2\pi \left(\frac{k}{M}m + \frac{l}{N}n\right)}$$

- Because the transform kernels are separable and symmetric, the two dimensional transforms can be computed as sequential row and column one-dimensional transforms.
- The basis functions of the transform are complex exponentials that may be decomposed into sine and cosine components.

TABLE 4.1

Summary of some important properties of the 2-D Fourier transform.

Property	Expression(s)
Fourier transform	$F(u,v) = \frac{1}{MN} \sum_{x=0}^{M-1} \sum_{y=0}^{N-1} f(x,y) e^{-j2\pi(ux/M + vy/N)}$
Inverse Fourier transform	$f(x, y) = \sum_{u=0}^{M-1} \sum_{v=0}^{N-1} F(u, v) e^{j2\pi(ux/M + vy/N)}$
Polar representation	$F(u,v) = F(u,v) e^{-j\phi(u,v)}$
Spectrum	$ F(u,v) = [R^2(u,v) + I^2(u,v)]^{1/2}, R = \text{Real}(F) \text{ and } I = \text{Imag}(F)$
Phase angle	$\phi(u,v) = \tan^{-1} \left[\frac{I(u,v)}{R(u,v)} \right]$
Power spectrum	$P(u,v) = F(u,v) ^2$
Average value	$\overline{f}(x, y) = F(0, 0) = \frac{1}{MN} \sum_{x=0}^{M-1} \sum_{y=0}^{N-1} f(x, y)$
Translation	$f(x,y)e^{j2\pi(u_0x/M+v_0y/N)} \Leftrightarrow F(u-u_0,v-v_0)$
	$f(x-x_0, y-y_0) \Leftrightarrow F(u,v)e^{-j2\pi(ux_0/M+vy_0/N)}$
	When $x_0 = u_0 = M/2$ and $y_0 = v_0 = N/2$, then
	$f(x, y)(-1)^{x+y} \Leftrightarrow F(u - M/2, v - N/2)$
	$f(x - M/2, y - N/2) \Leftrightarrow F(u, v)(-1)^{u+v}$

Conjugate symmetry
$$F(u,v) = F^*(-u,-v)$$
 $|F(u,v)| = |F(-u,-v)|$

Differentiation $\frac{\partial^n f(x,y)}{\partial x^n} \Leftrightarrow (ju)^n F(u,v)$ $(-jx)^n f(x,y) \Leftrightarrow \frac{\partial^n F(u,v)}{\partial u^n}$

Laplacian $\nabla^2 f(x,y) \Leftrightarrow -(u^2+v^2)F(u,v)$

Distributivity $\Im[f_1(x,y)+f_2(x,y)] = \Im[f_1(x,y)] + \Im[f_2(x,y)]$ $\Im[f_1(x,y)+f_2(x,y)] \neq \Im[f_1(x,y)] \cdot \Im[f_2(x,y)]$

Scaling $af(x,y) \Leftrightarrow aF(u,v), f(ax,by) \Leftrightarrow \frac{1}{|ab|}F(u/a,v/b)$

Rotation $x = r\cos\theta \quad y = r\sin\theta \quad u = \omega\cos\varphi \quad v = \omega\sin\varphi$ $f(r,\theta+\theta_0) \Leftrightarrow F(\omega,\varphi+\theta_0)$

Periodicity $F(u,v) = F(u+M,v) = F(u,v+N) = F(u+M,v+N)$ $f(x,y) = f(x+M,y) = f(x,y+N) = f(x+M,y+N)$

Separability See Eqs. (4.6-14) and (4.6-15). Separability implies that we can compute the 2-D transform of an image by first computing 1-D transforms along each row of the image, and then computing a 1-D transform along each column of this intermediate result. The reverse, columns and then rows, yields the same result.

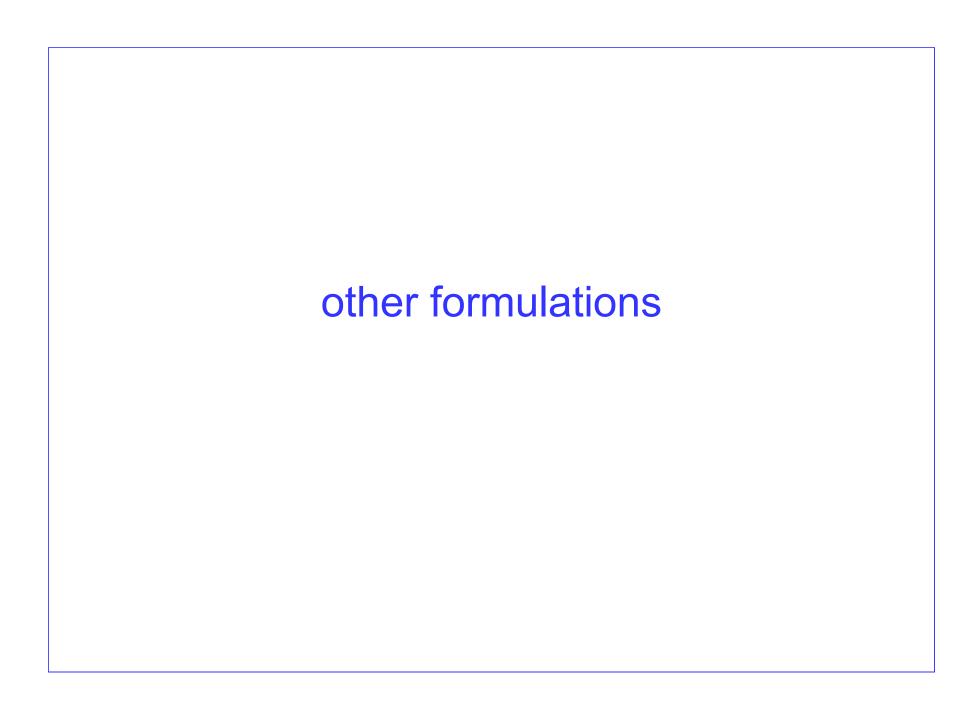
Property	Expression(s)
Computation of the inverse Fourier transform using a forward transform algorithm	$\frac{1}{MN}f^*(x,y) = \frac{1}{MN}\sum_{u=0}^{M-1}\sum_{v=0}^{N-1}F^*(u,v)e^{-j2\pi(ux/M+vy/N)}$ This equation indicates that inputting the function $F^*(u,v)$ into an algorithm designed to compute the forward transform (right side of the preceding equation) yields $f^*(x,y)/MN$. Taking the complex conjugate and multiplying this result by MN gives the desired inverse.
Convolution [†]	$f(x,y) * h(x,y) = \frac{1}{MN} \sum_{m=0}^{M-1} \sum_{n=0}^{N-1} f(m,n)h(x-m,y-n)$
Correlation [†]	$f(x, y) \circ h(x, y) = \frac{1}{MN} \sum_{m=0}^{M-1} \sum_{n=0}^{N-1} f^*(m, n) h(x + m, y + n)$
Convolution theorem [†]	$f(x, y) * h(x, y) \Leftrightarrow F(u, v)H(u, v);$ $f(x, y)h(x, y) \Leftrightarrow F(u, v) * H(u, v)$
Correlation theorem [†]	$f(x, y) \circ h(x, y) \Leftrightarrow F^*(u, v)H(u, v);$ $f^*(x, y)h(x, y) \Leftrightarrow F(u, v) \circ H(u, v)$

Some useful FT	pairs:
Impulse	$\delta(x,y) \Leftrightarrow 1$
Gaussian	$A\sqrt{2\pi}\sigma e^{-2\pi^2\sigma^2(x^2+y^2)} \Leftrightarrow Ae^{-(u^2+v^2)/2\sigma^2}$
Rectangle	$rect[a,b] \Leftrightarrow ab \frac{\sin(\pi ua)}{(\pi ua)} \frac{\sin(\pi vb)}{(\pi vb)} e^{-j\pi(ua+vb)}$
Cosine	$\cos(2\pi u_0 x + 2\pi v_0 y) \Leftrightarrow$
	$\frac{1}{2} \big[\delta(u + u_0, v + v_0) + \delta(u - u_0, v - v_0) \big]$

Sine
$$\sin(2\pi u_0 x + 2\pi v_0 y) \Leftrightarrow$$

 $j\frac{1}{2} [\delta(u + u_0, v + v_0) - \delta(u - u_0, v - v_0)]$

[†] Assumes that functions have been extended by zero padding.



2D Discrete Fourier Transform (DFT)

$$F[k,l] = \underbrace{\frac{1}{MN} \sum_{m=0}^{N-1} \sum_{n=0}^{N-1} f[m,n] e^{-j2\pi \left(\frac{k}{M}m + \frac{l}{N}n\right)}}_{-j2\pi \left(\frac{k}{M}m + \frac{l}{N}n\right)}$$
 where $l = 0,1,...,N-1$ $k = 0,1,...,M-1$

Inverse DFT

$$f[m,n] = \sum_{k=0}^{M-1} \sum_{l=0}^{N-1} F[k,l] e^{j2\pi \left(\frac{k}{M}m + \frac{l}{N}n\right)}$$

It is also possible to define DFT as follows

$$F[k,l] = \underbrace{\frac{1}{\sqrt{MN}} \sum_{m=0}^{M-1} \sum_{n=0}^{N-1} f[m,n] e^{-j2\pi \left(\frac{k}{M}m + \frac{l}{N}n\right)}}_{-j2\pi \left(\frac{k}{M}m + \frac{l}{N}n\right)}$$
 where
$$k = 0,1,...,M-1$$

$$l = 0,1,...,N-1$$

Inverse DFT

$$f[m,n] = \frac{1}{\sqrt{MN}} \sum_{k=0}^{M-1} \sum_{l=0}^{N-1} F[k,l] e^{j2\pi \left(\frac{k}{M}m + \frac{l}{N}n\right)}$$

Or, as follows

$$F[k,l] = \sum_{m=0}^{M-1} \sum_{n=0}^{N-1} f[m,n] e^{-j2\pi \left(\frac{k}{M}m + \frac{l}{N}n\right)}$$

where k = 0, 1, ..., M - 1 and l = 0, 1, ..., N - 1

Inverse DFT

$$f[m,n] = \frac{1}{MN} \sum_{l=0}^{M-1} \sum_{l=0}^{N-1} F[k,l] e^{j2\pi \left(\frac{k}{M}m + \frac{l}{N}n\right)}$$

2D DFT

 The discrete two-dimensional Fourier transform of an image array is defined in series form as

$$\mathcal{F}(u, v) = \underbrace{\frac{1}{N}}_{j=0}^{N-1} \sum_{k=0}^{N-1} F(j, k) \exp\left\{\frac{-2\pi i}{N} (uj + vk)\right\}$$

inverse transform

$$F(j,k) = \underbrace{\frac{1}{N}}_{u=0}^{N-1} \sum_{v=0}^{N-1} \mathcal{F}(u,v) \exp\left\{\frac{2\pi i}{N}(uj+vk)\right\}$$

2D DCT Discrete Cosine Transform

2D DCT

based on most common form for 1D DCT

$$C(u) = \alpha(u) \sum_{x=0}^{N-1} f(x) \cos \left[\frac{\pi (2x+1)u}{2N} \right],$$

$$f(x) = \sum_{u=0}^{N-1} \alpha(u)C(u)\cos\left[\frac{\pi(2x+1)u}{2N}\right],$$

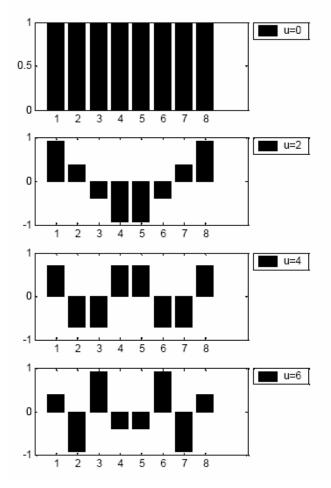
$$\alpha(u) = \begin{cases} \sqrt{\frac{1}{N}} & for \quad u = 0\\ \sqrt{\frac{2}{N}} & for \quad u \neq 0. \end{cases}$$

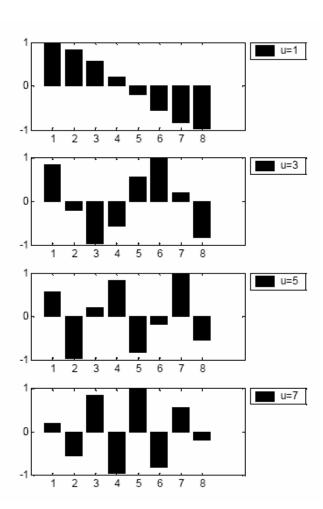
$$C(u = 0) = \sqrt{\frac{1}{N}} \sum_{x=0}^{N-1} f(x).$$

"mean" value

1D basis functions







Cosine basis functions are orthogonal

2D DCT

Corresponding 2D formulation

direct
$$C(u,v) = \alpha(u)\alpha(v) \sum_{x=0}^{N-1} \sum_{y=0}^{N-1} f(x,y) \cos\left[\frac{\pi(2x+1)u}{2N}\right] \cos\left[\frac{\pi(2y+1)v}{2N}\right],$$

$$u,v=0,1,....,N-1$$

$$\alpha(u) = \begin{cases} \sqrt{\frac{1}{N}} & for \quad u = 0\\ \sqrt{\frac{2}{N}} & for \quad u \neq 0. \end{cases}$$

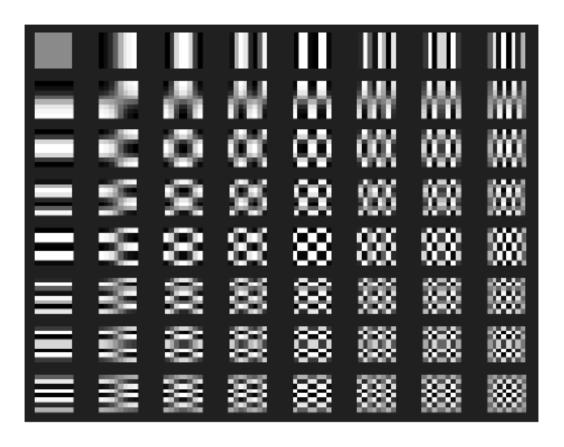
inverse
$$f(x, y) = \sum_{u=0}^{N-1} \sum_{v=0}^{N-1} \alpha(u)\alpha(v)C(u, v)\cos\left[\frac{\pi(2x+1)u}{2N}\right]\cos\left[\frac{\pi(2y+1)v}{2N}\right],$$

2D basis functions

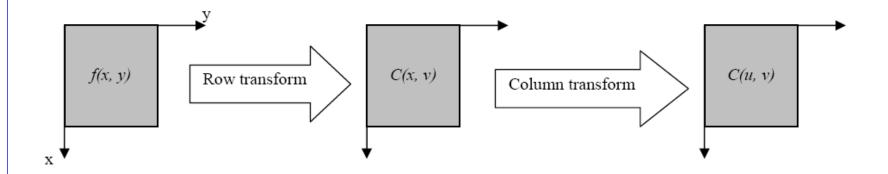
- The 2-D basis functions can be generated by multiplying the horizontally oriented 1-D basis functions (shown in Figure 1) with vertically oriented set of the same functions.
- The basis functions for N = 8 are shown in Figure 2.
 - The basis functions exhibit a progressive increase in frequency both in the vertical and horizontal direction.
 - The top left basis function assumes a constant value and is referred to as the DC coefficient.

2D DCT basis functions

Figure 2



Separability



The inverse of a multi-dimensional DCT is just a separable product of the inverse(s) of the corresponding one-dimensional DCT , e.g. the one-dimensional inverses applied along one dimension at a time

Separability

Symmetry

- Another look at the row and column operations reveals that these operations are functionally identical. Such a transformation is called a symmetric transformation.
- A separable and symmetric transform can be expressed in the form

$$T = AfA$$

where A is a NxN symmetric transformation matrix which entries a(i,j) are given by

$$a(i,j) = \alpha(j) \sum_{j=0}^{N-1} \cos \left[\frac{\pi(2j+1)i}{2N} \right],$$

 This is an extremely useful property since it implies that the transformation matrix can be pre computed offline and then applied to the image thereby providing orders of magnitude improvement in computation efficiency.

Computational efficiency

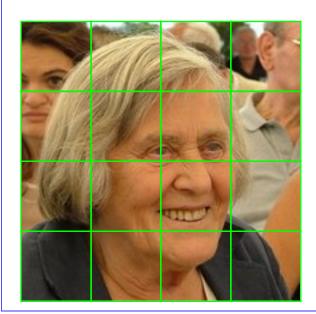
- Computational efficiency
 - Inverse transform $f = A^{-1}TA^{-1}$.
 - DCT basis functions are orthogonal. Thus, the inverse transformation matrix of A is equal to its transpose i.e. $A^{-1} = A^{T}$. This property renders some reduction in the pre-computation complexity.

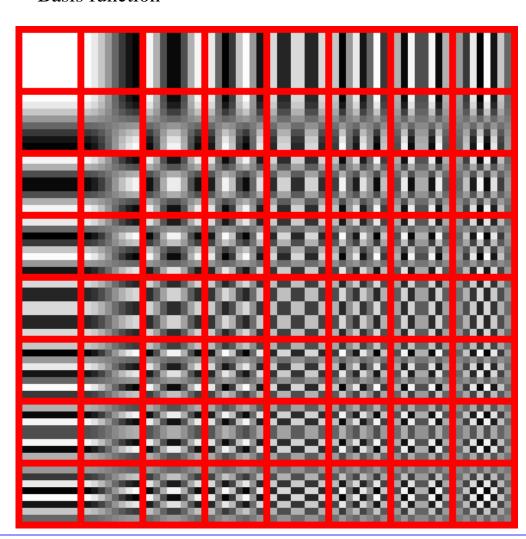
Block-based implementation Basis function

Block-based transform

Block size N=M=8

The source data (8x8) is transformed to a linear combination of these 64 frequency squares.





Energy compaction

(a)

(b)











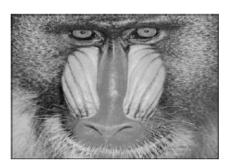


Energy compaction



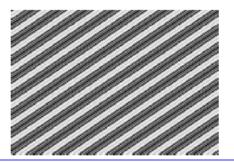








(e)





Appendix

Eulero's formula

$$A(j,k;u,v) = \exp\left\{\frac{-2\pi i}{N}(uj+vk)\right\} = \cos\left\{\frac{2\pi}{N}(uj+vk)\right\} - i\sin\left\{\frac{2\pi}{N}(uj+vk)\right\}$$

$$B(j,k;u,v) = \exp\left\{\frac{2\pi i}{N}(uj+vk)\right\} = \cos\left\{\frac{2\pi}{N}(uj+vk)\right\} + i\sin\left\{\frac{2\pi}{N}(uj+vk)\right\}$$